基于输出反馈的建筑结构闭开环次优控制。

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摘要:对传统的结构抗震闭开环控制算法进行改进。基于地面运动自回归模型,采用 Kalman 滤波 利用可以量测到的地面加速度激励对未来时段即将发生的地面加速度激励进行预估,并在微分方 程的求解中引入精确高效的精细积分算法。考虑到实际控制中量测全部状态变量的困难,改进算 法仅需量测部分状态变量。数值仿真表明,基于输出反馈的闭开环次优控制策略能大大降低结构 的地震响应。

关键词:地震激励;输出反馈;闭开环控制;Kalman 滤波;精细积分
 中图分类号:TU352.1
 文献标志码:A
 文章编号:1000-0844(2015)04-0933-05
 DOI:10.3969/j.issn.1000-0844.2015.04.0933

Closed/Open-loop Sub-optimal Control of Structures Based on Output Feedbacks

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Abstract: Most recent studies have been based on the application of linear quadratic regulator control to earthquake-excited structures. In linear quadratic regulator control problems, the objective function is defined as the integral of a quadratic expression in the control interval with respect to structural states and control vectors, and the optimal regulator can be derived using Pontryagin' s maximum principle or Bellman's method of dynamic programming. In traditional linear quadratic regulator control problems, the Riccati equation is obtained without considering the earthquake excitation term. To optimize control and satisfy the optimality condition, in this study, we propose a new closed/open-loop control strategy for structures under earthquake excitation. We derive an analytical solution to a linear regulator problem for structural control without neglecting unknown disturbances. The optimal regulator depends on both the state and disturbances. The solution for this closed/open-loop control requires the knowledge of the earthquake in the control interval, which is approximated based on the real-time prediction of near-future earthquake excitation using the Kalman filtering technique. Earthquake excitation is modeled as an autoregressive process. The prediction algorithm can predict seismic excitation in the near future with high accuracy, although it lacks prediction accuracy for more distant future events. Considering the measurement difficulty of all state variables, especially for some high-order systems, the proposed control strategy only requires the measurement of a partial state. In the calculation of a state tran-

① 收稿日期:2014-08-20

基金项目:交通运输部应用基础研究项目(2013319740080,2014319740160);交通运输部信息化技术研究项目(2013364740600);重 庆市科技人才培养计划项目(cstc2013kjrc-qnrc30001)

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sition matrix, which is required to solve a differential equation, large rounding errors may occur when the time-step size is excessively small. To overcome this limitation, we introduce a precise integration algorithm to solve the differential equation. This algorithm is always numerically stable and yields very high precision solutions for numerical integration problems. To demonstrate the effectiveness of the proposed control strategy, we investigated the undamped vibration of a three-story building subjected to horizontal seismic forces. We assumed that the columns of the building are massless and that the mass of the structure is concentrated at floor levels. We implemented control using actuators exerting forces on each story. We also assumed that floor velocities can be measured in real time by sensors installed in every story unit. We used the NS component of the 1940 El Centro earthquake ground acceleration record as the excitation source and performed calculations for its entire duration. We modeled the columns of the building as linear elastic springs and assumed the response mitigation effect of the actuators to be sufficient for the building to behave in a linear elastic manner during earthquake excitation. We did not consider the soil-structure interaction or the dynamic characteristics of the actuators. We investigated the controlled and uncontrolled behavior of the three-story undamped building and compared the relative displacement, velocity, acceleration, and inter-story displacement responses. Our numerical simulation results show that the proposed closed/open-loop sub-optimal output feedback control strategy can significantly reduce structural earthquake responses.

Key words: earthquake excitation; output feedback; close/open-loop control; Kalman filter; precise integration

0 引言

近几十年来控制在理论和实际工程中都获得了 很大进展^[1-3],其中线性二次型调节器(LQR)在很 多工程中得到了应用[4-5]。在传统的二次型调节器 问题中,目标函数定义为由结构状态和控制力向量 组成的二次表达式在一定时间区段上的积分,通过 庞德里亚金极大值原理或贝尔曼动态规划等方法可 推导出相应的最优控制器。然而,在公式推导过程 中 Riccati 方程是通过忽略外激励项而得出的。因 此从这种意义上说,传统的二次型闭环最优控制只 是一种近似的最优控制,在公式推导过程中保留外 激励项的闭开环控制比闭环控制有一定的优越性。 但是传统的闭开环控制需要预先知道整个控制时间 区段上的外激励,这对结构抗震等工程问题来说是 无法实现的。地震激励是随机的,人们无法事先知 道作用在结构上的确切地震激励。相应地不需要事 先知道作用在结构上的确切外激励的闭开环次优控 制方法应运而生。文献[6]基于地震自回归模型,通 过 Kalman 滤波预测一步或多步的地震动输入,采 用 Taylor 级数进行数值积分,提出一种次优的结构 抗震闭开环控制策略,并通过数值算例验证了该闭 开环次优控制方法相对于其他控制方法的优点。文 献[7]进一步提出具有指定稳定度的结构闭开环次 优控制算法。除此之外,为了避免 Riccati 方程的求 解和确定未来时段的地震激励,文献[8]提出一种基 于多点瞬态激励的闭环控制。为了提高控制精度, 文献[9]在结构瞬态闭环及瞬态闭开环控制中采用 了精细积分算法。

然而,以上文献[6-9]都假定结构的全部状态可 以量测,这在实际控制中往往是难以做到的。钟万 勰近年来提出的精细积分法求解常微分方程精度之 高^[10-12],是其他时域积分法无法比拟的。本文在文 献[6-9]的基础上,提出基于输出反馈的结构抗震闭 开环次优控制策略,对微分方程的求解采用精细积 分法来代替文献[6]和[7]中所采用的 Taylor 级数 展开法。

1 问题描述

以单维地震动输入下的 n 自由度线性结构为 例。假定地面均匀一致运动,结构的运动方程可写 为:

 $\boldsymbol{M}\ddot{\boldsymbol{x}}(t) + \boldsymbol{C}\dot{\boldsymbol{x}}(t) + \boldsymbol{K}\boldsymbol{x}(t) = -\boldsymbol{M}\boldsymbol{E}\ddot{\boldsymbol{x}}_{g}(t) + \boldsymbol{L}\boldsymbol{u}(t)$ (1)

其中:M、C、K分别为结构的质量阵、阻尼阵和刚度 阵;x(t)是相对地面的结构位移向量;E是所有分 量都为1的向量;u(t)为控制力向量;L为作用器位 置矩阵; $\ddot{x}_{g}(t)$ 为地面加速度。式(1)可写为状态空间形式

$$\dot{z}(t) = \mathbf{A}z(t) + \mathbf{B}u(t) + \mathbf{H}\ddot{\mathbf{x}}_{g}$$
(2)

其中:

$$A = \begin{bmatrix} 0 & I \\ -M^{-1}K & -M^{-1}C \end{bmatrix}, B = \begin{bmatrix} 0 \\ M^{-1}L \end{bmatrix},$$
$$H = \begin{bmatrix} 0 \\ -E \end{bmatrix}, z = \begin{cases} x \\ \dot{x} \end{cases}$$

1.1 基于状态反馈的闭环控制

采用性能指标

$$J = \frac{1}{2} \int_{0}^{t_f} (\boldsymbol{z}^T \boldsymbol{Q} \boldsymbol{z} + \boldsymbol{u}^T \boldsymbol{R} \boldsymbol{u}) dt$$
 (3)

其中:t_f 表示地震激励持续时间。引入哈密顿函数

$$H_{a} = \frac{1}{2} \left[\boldsymbol{z}^{T}(t) \boldsymbol{Q} \boldsymbol{z}(t) + \boldsymbol{u}^{T} \boldsymbol{R} \boldsymbol{u}(t) \right] + \boldsymbol{\lambda}^{T}(t) \left[\boldsymbol{A} \boldsymbol{z}(t) + \boldsymbol{B} \boldsymbol{u}(t) + \boldsymbol{H} \ddot{\boldsymbol{x}}_{g}(t) \right]$$
(4)

由方程

$$\dot{z}(t) = \frac{\partial H_a}{\partial \lambda} = Az(t) + Bu(t) + H\ddot{x}_g(t) \quad (5)$$

$$\dot{\boldsymbol{\lambda}}(t) = -\frac{\partial H_a}{\partial \boldsymbol{z}} = -\boldsymbol{Q}\boldsymbol{z}(t) - \boldsymbol{A}^{\mathrm{T}}\boldsymbol{\lambda}(t) \qquad (6)$$

$$\frac{\partial H_a}{\partial \boldsymbol{u}} = \boldsymbol{0} = \boldsymbol{R}\boldsymbol{u}(t) + \boldsymbol{B}^{\mathrm{T}}\boldsymbol{\lambda}(t)$$
(7)

$$\boldsymbol{\lambda}(t_f) = \boldsymbol{0} \tag{8}$$

假定

$$\boldsymbol{\lambda}(t) = \boldsymbol{P}(t)\boldsymbol{z}(t) \tag{9}$$

联立式(5)~式(9),可得

$$\begin{bmatrix} \dot{\boldsymbol{P}}(t) + \boldsymbol{P}(t)\boldsymbol{A}(t) + \boldsymbol{A}^{T}(t)\boldsymbol{P}(t) - \boldsymbol{P}(t)\boldsymbol{B}\boldsymbol{R}^{-1}\boldsymbol{B}^{T}\boldsymbol{P}(t) + \\ \boldsymbol{Q} \end{bmatrix} \boldsymbol{z}(t) + \boldsymbol{P}(t)\boldsymbol{H}\ddot{\boldsymbol{x}}_{g}(t) = \boldsymbol{0}; \boldsymbol{P}(t_{f}) = \boldsymbol{0} \quad (10)$$

式(10)对所有的z(t)都应成立。忽略激励项 \ddot{x}_{g} (t),式(10)可另写为

$$\dot{\boldsymbol{P}}(t) + \boldsymbol{P}(t)\boldsymbol{A}(t) + \boldsymbol{A}^{\mathrm{T}}(t)\boldsymbol{P}(t) - \boldsymbol{P}(t)\boldsymbol{B}\boldsymbol{R}^{-1}\boldsymbol{B}^{\mathrm{T}}\boldsymbol{P}(t) + \boldsymbol{Q} = \boldsymbol{0};$$
$$\boldsymbol{P}(t_{f}) = \boldsymbol{0}$$
(11)

联立式(7)和(9),可得

$$\boldsymbol{u}(t) = -\boldsymbol{R}^{-1}\boldsymbol{B}^{\mathrm{T}}\boldsymbol{P}(t)\boldsymbol{z}(t) \qquad (12)$$

由式(11)解得 **P**(*t*),代入式(12),便得到 *t* 时刻作 用器应输出的控制力。

1.2 基于状态反馈的闭开环控制

不忽略激励项 $\ddot{x}_{g}(t)$,假定

$$\boldsymbol{\lambda}(t) = \boldsymbol{P}(t)\boldsymbol{z}(t) + \boldsymbol{q}(t)$$
(13)
6) (7) fu(12) at 4

联立式(5)、(6)、(7)和(13),可得

$$\dot{\boldsymbol{P}}(t) + \boldsymbol{P}(t)\boldsymbol{A}(t) + \boldsymbol{A}^{T}(t)\boldsymbol{P}(t) - \boldsymbol{P}(t)\boldsymbol{B}\boldsymbol{R}^{-1}\boldsymbol{B}^{T}\boldsymbol{P}(t) +$$

$$\boldsymbol{\mathcal{Q}} \,] \,\boldsymbol{z}(t) + \dot{\boldsymbol{q}}(t) - \left[\boldsymbol{P}(t) \boldsymbol{B} \boldsymbol{R}^{-1} \boldsymbol{B}^{T} - \boldsymbol{A}^{T} \right] \boldsymbol{q}(t) + \\ \boldsymbol{P}(t) \boldsymbol{H} \ddot{\boldsymbol{x}}_{g}(t) = \boldsymbol{0}$$
(14)

式(14)在任何时刻均成立。联立式(8)和式(14),可得

$$\dot{\boldsymbol{P}}(t) + \boldsymbol{P}(t)\boldsymbol{A}(t) + \boldsymbol{A}^{\mathrm{T}}(t)\boldsymbol{P}(t) - \boldsymbol{P}(t)\boldsymbol{B}\boldsymbol{R}^{-1}\boldsymbol{B}^{\mathrm{T}}\boldsymbol{P}(t) + \boldsymbol{Q} = \boldsymbol{0};$$
$$\boldsymbol{P}(t_{f}) = \boldsymbol{0}$$
(15)

$$\dot{\boldsymbol{q}}(t) - \left[\boldsymbol{P}(t)\boldsymbol{B}\boldsymbol{R}^{-1}\boldsymbol{B}^{T} - \boldsymbol{A}^{T}\right]\boldsymbol{q}(t) + \boldsymbol{P}(t)\boldsymbol{H}\ddot{\boldsymbol{x}}_{g}(t) = \boldsymbol{0};$$
$$\boldsymbol{q}(t) = \boldsymbol{0}$$
(16)

联立式(7)和(13),可得

$$\boldsymbol{u}(t) = -\boldsymbol{R}^{-1}\boldsymbol{B}^{T}\left[\boldsymbol{p}(t)\boldsymbol{z}(t) + \boldsymbol{q}(t)\right] \quad (17)$$

若整个控制时间区段作用在结构上的地面加速 度激励 $\ddot{x}_{g}(t)$ 事先已知,由式(15)解得 P(t),由式 (16)解得 q(t),代入式(17)便得到 t 时刻作用器应 输出的控制力。

2 基于输出反馈的闭开环次优控制

式(15)为微分 Riccati 方程,借助于钟万勰近年 来提出的精细算法^[10-12],微分 Riccati 方程可以得到 简便地解决。然而在除临近 t_f 的时间段外,P(t)在 大部分时间段都近似保持为常数矩阵,把式(15)简 化为代数 Riccati 方程来处理,见下式:

PA + A^TP - PBR⁻¹B^TP + Q = 0 (18)
式(16)的求解需要从终端时刻进行后向积分,
这需要事先知道整个控制时间区段作用在结构上的
地面加速度激励。虽然地面加速度激励可以在线量
测,但却无法事先知道。采用地面运动自回归模型
利用 Kalman 滤波技术对未来时段即将发生的地面
加速度激励进行预估^[6]。

对于每一时间步[t_{k-1}, t_k],式(16)的解可表示 为

$$\boldsymbol{q}(t_k) = \boldsymbol{T}(\boldsymbol{\eta})\boldsymbol{q}(t_{k-1}) + \boldsymbol{c}_k \qquad (19)$$

其中:

$$\eta = t_k - t_{k-1}, \boldsymbol{T}(\eta) = e^{\boldsymbol{p}_{\eta}}, \boldsymbol{c}_k = \int_{t_{k-1}}^{t_k} e^{\boldsymbol{p}(t_k - t_d)} \boldsymbol{p}(t_d) dt_d$$

 $\boldsymbol{D} = \boldsymbol{P}\boldsymbol{B}\boldsymbol{R}^{-1}\boldsymbol{B}^{T} - \boldsymbol{A}^{T}, \ \boldsymbol{p}(t) = -\boldsymbol{P}\boldsymbol{H}\ddot{\boldsymbol{x}}_{g}(t)$

假定 $\ddot{\mathbf{x}}_{g}(t)$ 在时间步 $[t_{k-1}, t_{k}]$ 内线性变化,则

$$\boldsymbol{c}_{k} = \boldsymbol{T}(\boldsymbol{\eta})\boldsymbol{D}^{-1}\left[\boldsymbol{p}(t_{k-1}) + \boldsymbol{D}^{-1}\boldsymbol{r}_{1}\right] -$$

 $D^{-1} \left[\boldsymbol{p}(t_{k-1}) + \boldsymbol{D}^{-1} \boldsymbol{r}_1 + \boldsymbol{r}_1 \boldsymbol{\eta} \right]$ (20)

其中: $r_1 = [p(t_k) - p(t_{k-1})]/\eta$ 。而 $T(\eta)$ 的精细计 算可见文献[9,12]。通过递推, $q(t_k)$ 可以通过q(t)在初始时刻的值q(0)来表示:

$$\boldsymbol{q}(t_k) = \boldsymbol{T}^k(\boldsymbol{\eta})\boldsymbol{q}(0) + \sum_{i=1}^k \boldsymbol{T}^{k-i}(\boldsymbol{\eta})\boldsymbol{c}_i \quad (21)$$

记从初始时刻到终端时刻 t_f 的积分步数为m, 由 $q(t_f)=0$,可得

$$\boldsymbol{q}(0) = -\sum_{i=1}^{m} \boldsymbol{T}^{-i}(\boldsymbol{\eta}) \boldsymbol{c}_{i}$$
(22)

$$\boldsymbol{q}(t_{k}) = -\sum_{i=k+1}^{m} \boldsymbol{T}^{k-i}(\boldsymbol{\eta})\boldsymbol{c}_{i} = -\boldsymbol{T}^{-1}(\boldsymbol{\eta})\boldsymbol{c}_{k+1} - \boldsymbol{T}^{-2}(\boldsymbol{\eta})\boldsymbol{c}_{k+2} - \boldsymbol{\Lambda} - \boldsymbol{T}^{k-m}(\boldsymbol{\eta})\boldsymbol{c}_{m} \quad (23)$$

利用地面运动自回归模型和 Kalman 滤波技术 进行地面加速度激励 l 步预估,即在 t_k 时刻利用过 去时段(包括现在时刻)量测到的地面加速度激励估 计未来时段 $t_{k+1}, t_{k+2}, \dots, t_{k+l}$ (1 $\leq l \leq m-k$)时刻 的地面加速度激励。将式(23)截取前 l 项,可得 $q(t_k) \approx -T^{-1}(\eta) c_{k+1} - T^{-2}(\eta) c_{k+2} - \Lambda - T^{k-1}(\eta) c_{k+l}$ (24)

关于截断所带来的误差的讨论,可见文献[6-7]。式(17)需要量测结构全部状态变量,对于高阶 系统,这在很多情况下是不可实现的。这里假定只 有部分状态变量可测,记为 y(t):

$$\mathbf{y}(t) = \mathbf{C}_{\mathbf{y}} \mathbf{z}(t) \tag{25}$$

假定

$$\boldsymbol{u}(t) = -\boldsymbol{K}_{c}\boldsymbol{y}(t) - \boldsymbol{R}^{-1}\boldsymbol{B}^{T}\boldsymbol{q}(t) \qquad (26)$$

定义目标函数

$$J_2 = \| \boldsymbol{G} \|_{\mathrm{F}}^2 = \operatorname{trace}(\boldsymbol{G}^{\mathrm{T}}\boldsymbol{G}) \qquad (27)$$

其中: $G = K_{c}C_{y} - R^{-1}B^{T}P$; F 表示 Frobenius 范数; K_{c} 的取值应使 J_{2} 最小。令 $\frac{dJ_{2}}{dK_{1}} = 0$,可得

$$\boldsymbol{K}_{c} = \boldsymbol{R}^{-1} \boldsymbol{B} \boldsymbol{P} \boldsymbol{C}_{v}^{T} (\boldsymbol{C}_{v} \boldsymbol{C}_{v}^{T})^{-1}$$
(28)

将式(24)和(28)代入式(26)中,就可得到基于 输出反馈的结构闭开环次优控制。相应地忽略式 (26)中的 $R^{-1}B^{T}q(t)$,将式(28)代入式(26),即得 到基于输出反馈的结构闭环次优控制。

3 算例

一个三层无阻尼结构受水平方向地面加速度激励,相邻层间安装主动作用器,如图 1 所示^[6]。各层质量均为 48×10³ kg,层间刚度分别为 k_1 =46 420 kN/m, k_2 =41 780 kN/m, k_3 =23 210 kN/m。地面加速度激励采用 El Centro 地震波南北分量,地震加速度峰值为 3.417 m/s²,时程记录间距为 0.02 s。假定结构速度可测,即 C_y =[0 I]。对未来时段地面加速度激励进行 3 步预估,采用基于输出反馈的闭开环控制策略对结构进行控制。取加权矩阵

$$\boldsymbol{Q} = \mu \begin{bmatrix} \boldsymbol{0} & \boldsymbol{0} \\ \boldsymbol{0} & \boldsymbol{I} \end{bmatrix}, \ \boldsymbol{R} = 1.0 \times 10^{-5} \boldsymbol{I}$$
(29)

其中: μ 为可调参数。不同的 μ 值对应不同的控制 力和控制效果,经试算,取 $\mu=1.0\times10^{8}$ 。



图 1 结构模型 Fig.1 Structural model

这里仅给出闭开环控制和无控时结构峰值响 应对比,见表1。关于闭开环控制和闭环控制控制 效果的详细比较,可见文献[6-7]。从表1可以看 出,施加控制力后,结构顶层的位移、速度、加速度峰 值响应和2~3层层间相对位移峰值响应都大大降 低。无控时,结构顶层位移峰值为12.7 cm,施加控 制力后结构顶层位移峰值降为1.1 cm。

表 1 控制前后结构响应对比

Table 1 Responses of structure with and without control

峰值响应	无控	闭开环控制
顶层位移/cm	12.7	1.1
2~3 层层间相对位移/cm	5.4	0.2
顶层速度/(cm • s ⁻¹)	174.5	10.3
顶层加速度/(m•s ⁻²)	26.3	3.6
顶层控制力/kN	-	159.1
底层控制力/kN	-	364.5

4 结论

针对实际控制中量测全部状态的困难,提出一 种基于输出反馈的结构闭开环次优控制算法,并对 一个地震激励下的三层无阻尼结构进行数值仿真。 结果表明,输出反馈闭开环次优控制可以很好地实 现控制效果。

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